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FOR

**A Crossbar Device With Reduced Parasitic Capacitive Loading
and Usage of Crossbar Devices in Reconfigurable Circuits**

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BACKGROUND OF THE INVENTION

1. Field of the Invention

10 The present invention relates to the field of integrated circuit (IC). More specifically, the present invention relates to crossbar devices, and their usage in reconfigurable circuits.

2. Background Information

15 Crossbar devices for programmatically connecting n inputs to m outputs, in general, are known in the art.

Figures 1a-1b show a basic implementation of a pass n-mos crossbar device known in the art. Input lines 100 are connectable to output lines 101 through switches 102. Each switch 102 comprises a n-mos pass transistor 103 with its source connected to one input line and its drain connected to one output line; and a memory element 104 controlling the gate of pass transistor 103. Connection between one input line and one output line is effectuated by applying a high voltage (by storing a 1 in the memory element) to the gate of the corresponding pass transistor; putting the pass transistor in a low resistance state between its source and drain. Output buffer 105 amplifies and regenerates the voltage level on the output line, restoring the pass transistor 103 threshold voltage (V_{th}) drop. The configuration of such a crossbar should connect only one input to one output,

otherwise it can possibly create a short circuit between two inputs connected to the same output. This type of crossbar causes problems at power up if the memory element is unknown, possibly connecting several inputs to a same output. Also, this type of crossbar needs $n \times m$ memory elements to realize a n inputs to m outputs crossbar.

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Figure 2 shows another prior art implementation (US PAT 6,289,494) having a more efficient structure. This type of crossbar needs $(n / 4) \times m$ memory elements **201** and m 2 to 4 decoders **202** to realize a n inputs to m outputs crossbar. For a large crossbar input number the penalty of the 2 to 4 decoder **202** is compensated 10 by the reduction in the number of memory elements required. One problem with this architecture is the capacitive loading of the input lines. To connect input line **203a** to output line **204** a 1 is programmed into memory element **201a**, and the decoder drives a 1 on its output **205b**. Therefore, input line **203b** is connected to capacitor **206**. Similarly, every fourth input of one column is connected to a capacitor **206**. 15 Note, the capacitive load of one input depends on the programming pattern of the other inputs, which could ends up with a high capacitive load on some input lines and a low capacitive load on some other input lines. Also, the capacitance **206** is pretty big because it represents the parasitic load of five n-mos drains/sources and the metal interconnections between these five n-mos drains/sources.

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Figure 3 shows another prior art implementation (US PAT 5.260.610). This type of crossbar also needs $(n / 2) \times m$ memory elements **301** plus m memory elements **302**. To connect input line **303a** to output line **304** we must program a 1 in memory element **301** and a 1 in memory element **302**. But, by programming a one in memory element **301**, input line **303b** is connected to capacitance **306**.

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Capacitance **306** is large because it represents the parasitic load of half of the pass transistor of one column plus the metal interconnection between them. If the

X1
crossbar has 32 inputs, then capacitance 306 includes the parasitic load of 16 n-mos drains/sources. Again, the capacitive loading of one input lines can vary dramatically with the programming pattern of the other inputs.

X2 5
In application where a significant number of crossbar are employed and interconnected, such as reconfigurable circuit applications, the input capacitive load variation of one crossbar input with respect to the programming pattern of its other inputs makes the timing optimization of high performance devices very difficult. Additionally, these and other prior art crossbar devices are found to consume more power and/or area than desired, as well as contributing to current swing.

10 Thus, a crossbar device and techniques of employment in reconfigurable circuit without at least some of these disadvantages are desired.

SUMMARY OF THE INVENTION

A crossbar device includes a first set of input conductor lines and a second set of output conductor lines. A plurality of chains of pass transistors are provided to selectively couple the input lines to the output lines in a reduced parasitic capacitive loading manner. Further, memory elements and decoder logic are provided to facilitate control of the selective coupling.

In one embodiment, each pass transistor chain comprises a first pass transistor with its source connected to one input line and its gate connected to a memory element, and a second pass transistor with its source connected to the drain of the first pass transistor, its gate connected to a decoder logic, and its drain connected to one output line. For each output line, the memory element selects a first group of inputs and the decoder selects one input out of the first group of inputs thereby establishing a connection between the selected input line and the output line.

In accordance with another aspect, a low power application of multiple crossbar devices to a reconfigurable circuit block is improved by having each memory element of a crossbar device be provided with a supply voltage higher by a threshold voltage V_{th} to maintain the supply voltage of corresponding output buffers input at V_{dd} , to prevent the output buffers from consuming static current when their inputs are at a degenerated level, to facilitate the lower power application.

In accordance with yet another aspect, an application of multiple crossbar devices to a reconfigurable circuit block is improved by coupling a control circuitry via a control line to all output buffers of the interconnected crossbar devices to force all output buffers of the crossbar devices to a known state at power-on, to prevent unpredictable behavior.

BRIEF DESCRIPTION OF DRAWINGS

The present invention will be described by way of exemplary embodiments, but not limitations, illustrated in the accompanying drawings in which like references 5 denote similar elements, and in which:

Figures 1a-1b illustrate a basic crossbar implementation known in the art;

Figures 2-3 illustrate additional prior art implementations;

Figure 4 illustrates the improved crossbar device of the present invention, in accordance with one embodiment;

10 **Figure 5** illustrates the dual pass transistor layout, in accordance with one embodiment;

Figure 6 illustrates a low power application of crossbar devices in a reconfigurable circuit , in accordance with one embodiment; and

15 **Figure 7** illustrates a technique to improve employment of a large number of crossbar devices in a reconfigurable circuit, in accordance with one embodiment.

DETAILED DESCRIPTION OF THE INVENTION

In the following description, for purposes of explanation, specific numbers, materials and configurations are set forth in order to provide a thorough 5 understanding of the present invention. However, it will be apparent to one skilled in the art that the present invention may be practiced without the specific details. In other instances, well-known features are omitted or simplified in order not to obscure the present invention.

Further, the phrase "in one embodiment" is used repeatedly. In general, the 10 phrase does not refer to the same embodiment, although it may. The terms "comprising", "including", "having" and the like, as used in the present application, are synonymous.

Referring now to **Figures 4-5**, wherein the crossbar device of the present 15 invention, in accordance with one embodiment is shown. The crossbar device includes $(n / 4) \times m$ memory elements 401 plus m 2 to 4 decoders 402 for selectively control the coupling of n inputs to m outputs. As illustrated, the crossbar device advantageously uses a dual gate pass transistor structure to minimize the parasitic capacitive loading of the input lines. Each input line, e.g. input line 403a is 20 connected to a output line, such as output line 404, by way of a chain of pass transistors, such as pass transistors 407a and 407b.

To connect input line 403a to output line 404, a 1 is programmed in memory element 401 and the decoder 402 drives a 1 on line 405b. By programming a 1 in memory element 401 input line 403b is connected to capacitance 406. However, as 25 a result of the dual gate pass transistor structure, parasitic capacitance 406 is reduced to a small active area as depicted on **Figure 5**. Therefore, although the

input line parasitic load caused by the programming pattern of the other input lines is not totally suppressed, the present invention substantially reduces it to the layout of the drain/source area while employing a small number of memory elements to realize the crossbar device.

5 While for ease of understanding, **Fig. 4** illustrated the chain of pass transistors coupling an input line to an output line as having two pass transistors, in alternate embodiments, the present invention may be practiced with the chain as having more than two pass transistors with the first pass transistor, connected to the input line, controlled by a local memory element like **401** and the other pass 10 transistor of the chain controlled by a decoder circuitry or memory elements.

In low power applications where a number of crossbar devices are interconnected, it is important to reduce the voltage swing on the interconnections between the crossbar devices, to reduce the dynamic current. **Figure 6** shows a 15 low power structure suitable for use in a crossbar device, in accordance with another aspect of the present invention. The lower power attribute is achieved by lowering the supply voltage **601** of the crossbar output buffer **603** and the additional drivers providing inputs to the various crossbar of the circuit. When a VDD level is applied on input line **606**, output line **607** receives a degenerated level VDD – 20 threshold voltage V_{th} , because of the V_{th} drop across n-mos pass transistors **602**. This degenerated level when applied on the input of crossbar output buffer **603** produces a parasitic current flowing through the first inverter stage. This parasitic current, when multiplied by a significant number of crossbar output buffers of an integrated circuit block, such as a FPGA block, may ruin the low power target. As 25 illustrated in **Figure 6**, the V_{th} drop is advantageously compensated by raising supply voltage **605** for memory element **604** by V_{th} . As a result, the gate of the pass

transistor 602 receives a voltage level that is one V_{th} above the voltage level of input line 606. However, this voltage raise does not impact the circuit power because there is no static or dynamic current in the memory element during the circuit operation. Resultantly, the crossbar output line 607 has a full VDD level.

5 The crossbar device may be the crossbar device of the present invention as illustrated in **Fig 4-5**. Alternatively, the crossbar device may also be a crossbar device of the prior art having the requisite memory and output buffer elements.

As alluded to earlier, employing crossbar devices in a low power manner is especially desirable for an integrated circuit or integrated circuit block where a 10 significant number of crossbar devices are employed and interconnected. An example of such integrated circuit is the scalable reconfigurable circuit disclosed in co-pending U.S. Patent Application, number <to be inserted>, entitled "A 15 Reconfigurable Integrated Circuit Having a Scalable Architecture", filed <to be inserted>, having common inventorship with present application. The specification of which is hereby fully incorporated by reference.

The present invention contemplates IP blocks incorporated with the teachings of the present invention for incorporation into reconfigurable integrated circuits, as well as integrated circuits directly practicing the teachings of the present invention.

20 **Figure 7** shows an improved crossbar output buffer structure to avoid static current at power-up, in accordance with yet another aspect of the present invention. As mentioned above, at power-up, the state of the memory elements are undefined. This may create various paths between the inputs of a crossbar. For a 25 reconfigurable circuit block, such as the one disclosed in copending application '123, many output buffers may be shorted together at power-up, producing a large current flow through the device. Also, during configuration loading sequence of the circuit

block, the incomplete configuration may temporary create short circuits between the crossbar inputs. To compensate for these possibilities, the output buffers 704 are advantageously connected to a global control line 701 forcing their outputs to a known level. This control line is activated by a power-on reset circuitry 702 and is 5 deactivated when a configuration has been loaded in the reconfigurable circuit block. Since all the crossbar outputs are at the same level during the power-up and until a configuration is loaded, the fact that they may or may not be shorted together does not produce any more current. For example, at power up, power on reset circuitry 702 resets the flip-flop 703. The flip-flop output 701 forces all the crossbar 10 buffers 704 to zero. When a configuration is loaded, flip flop 702 is written with a logical 1, enabling all crossbar output buffers.

The crossbar device may be the crossbar device of the present invention as illustrated in **Fig 4-5**. Alternatively, the crossbar device may also be a crossbar device of the prior art having the requisite memory and output buffer elements. 15 Further, the technique disclosed in **Fig. 7** may also be practiced in conjunction with the voltage supply technique of **Fig. 6**.

As alluded to earlier, the present invention contemplates IP blocks incorporated with the teachings of the present invention for incorporation into reconfigurable integrated circuits, as well as integrated circuits directly practicing the 20 teachings of the present invention.

Thus an improved crossbar device with reduced parasitic capacitive loading, and improved techniques for using a significant number of crossbar devices have been disclosed. As alluded to earlier, the described embodiments are illustrative, 25 and not restrictive. The present invention may be practiced with modifications and

alterations to the described embodiments, consistent with the scope of the invention as set forth by the claims below.